

# **SDR Cube**



A Portable Software Defined Radio Utilizing An Embedded DSP Engine for Quadrature Sampling Transceivers

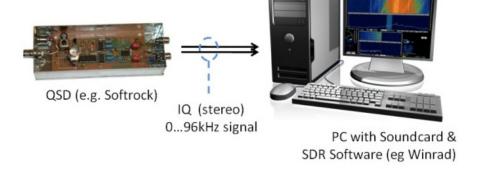
George Heron N2APB and Juha Niinikoski, OH2NLT

As presented at the 29th Annual TAPR Digital Communications Conference on Sept 25, 2010 in Vancouver, WA



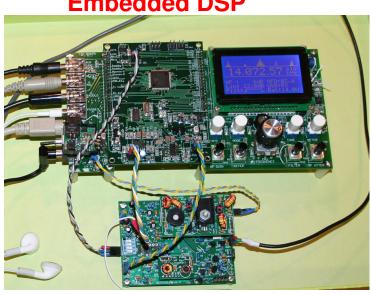
## What is the SDR Cube?

#### Instead of this ...



#### **Embedded DSP**

We do THIS ...



**Softrock RXTX 6.3** 

**Graphic User Interface** 

> Video clip 1 Video clip 2



### So In A Nutshell ...

## "We don't need no stinkin' PC!"

(Apologies to Alfonso Bedoya in the 1948 classic film "Treasure of the Sierra Madres";-)

Processing the I and Q channels of any Softrock-like RF "front end" with an **embedded computer** (dsPIC) provides ...

- Smaller package
- Easier to go portable
- Safer, worry-free operation in the field
- Can actually see the display in bright sunlight
- Longer lasting on batteries
- Lower cost

"Most of the benefits of SDR, without the hassles of lugging the laptop."



## **BACKGROUND**

"What's brought us to this point?"

- Holy Grail of untethering SDR from the PC, for portability, et al.
- Early "phasing method" literature by Frohne, Johnson, Campbell, Larkin, et al
- OH2NLT CheapDSP experiments a number of years ago
- Softrock popularity today → More than 11,000 kits in the field, thanks to Tony Parks, KB9YIG
- "Give me I and Q, and I'll conquer the world!"



# **DESIGN PARTNER ... Juha Niinikoski**

QTH: Espoo, Finland





## **CheapDSP**

The CheapDSP experimental project in 2005 was a learning exercise leading to a direct conversion transceiver developed by OH2NLT and OH7SV. Experiences from this project formed the foundation of the SDR Cube DSP

block.



Link to early experiments ...

http://www.kolumbus.fi/juha.niinikoski/Cheap\_dsp/Cheap\_dsp.htm



## **DEVELOPMENT PLATFORM**



The three boards constituting the SDR Cube were designed such that they could inter-connect in a compact manner in a 4" cube-sized cabinet, or side-by-side on the bench for convenient access and experimenting as shown above. (The Softrock RXTX 6.3 is shown below the Cube board set.)



## **MAJOR FEATURES**

- Standalone SDR transceiver → no PC, portable, compact
- Self-contained single band → based on the I/Q RF front end
- **Softrock-compatible** → designed to interface with SR v6.3 RXTX, etc.
- Low Power → 90ma (Cube), plus 100 ma (Softrock Rx) or 300 ma (Software Tx
- Add-on RF Amp & Attenuator → good control of incoming RF, correct some SR imperfections. (Sensitivity to antenna impedance variations.)
- Quadrature Sampling Clocking options → DDS, Si570, or I2C to target Softrock
- Built-in Keyer → 1-50 wpm, lambic A, B, or straight key
- Popular HF modes → SSB, CW, AM, Digital (with special interface to NUE-PSK)
- Special interface to NUE-PSK Modem → digital interface provides best quality
- Graphic LCD Display → Provides clear indications of the many status and options
- Bandscope → provides +/- 4 kHz spectrum visibility for Rx, signal monitor for Tx
- Audio filtering → low corner 200Hz, high corners 700, 1500, 2400 or 3600Hz
- Audio Output → Headphones or amplified speaker, Binaural Audio
- **Beeper** → UI clicks/feedback, code practice oscillator
- Frequency agility → Fast/Med/Slo tune, dual VFOs, memories, RIT/XIT
- **Menus** → Calibration, all settings, system gain, sidetone frequency,
- Software Upgradeable -> Bootloader enables user to load new software versions
- Open Source & Hardware



## HARDWARE ARCHITECTURE

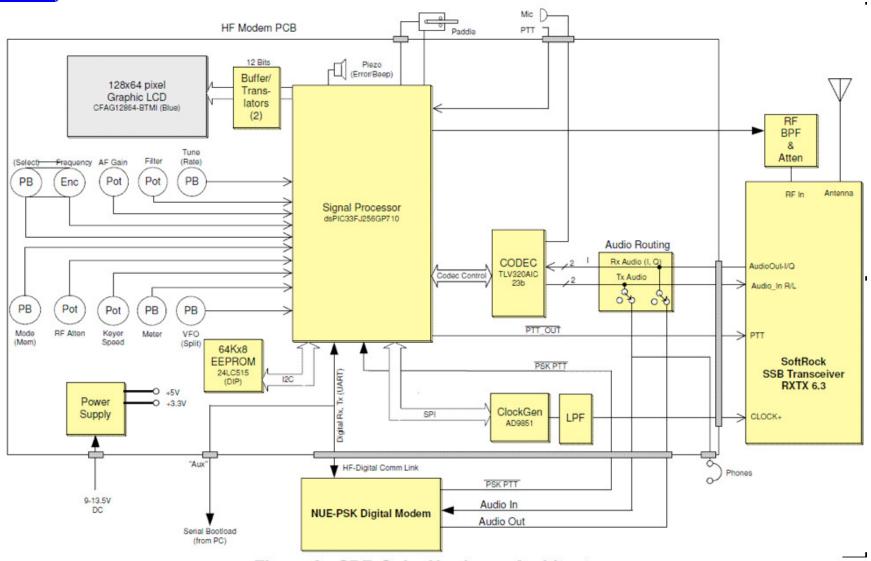


Figure 3: SDR Cube Hardware Architecture

Illustrates designed-in controls, graphic display, and tight integration with NUE-PSK Digital Modem



## SOFTWARE ARCHITECTURE

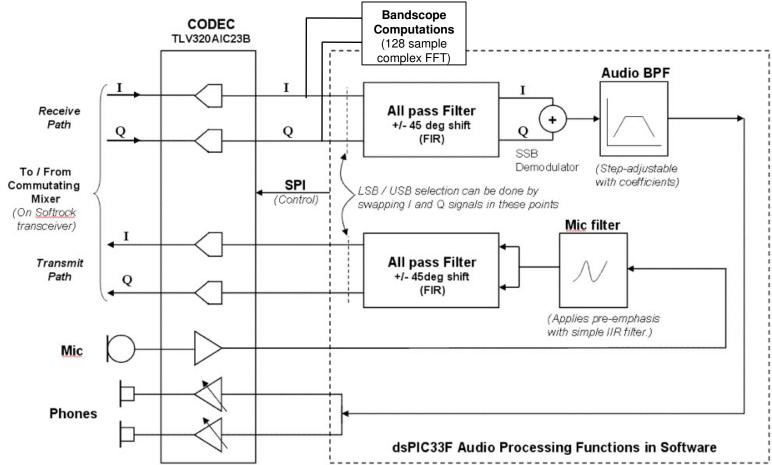
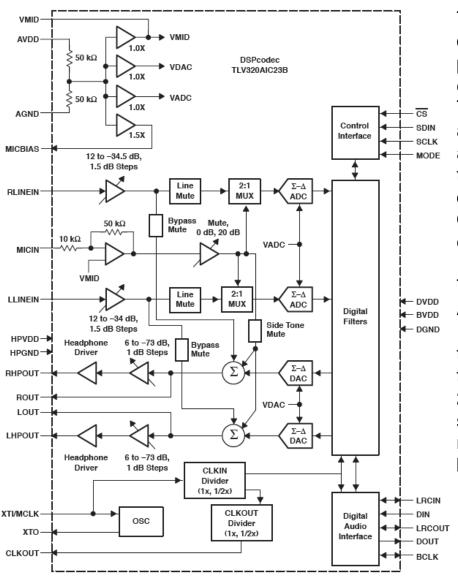


Figure 2: Software Processing Block Diagram

This is just a starting point. All transceiver functions are softwarebased in the DSP and they can be continuously improved.



#### TLV320AIC23B Codec

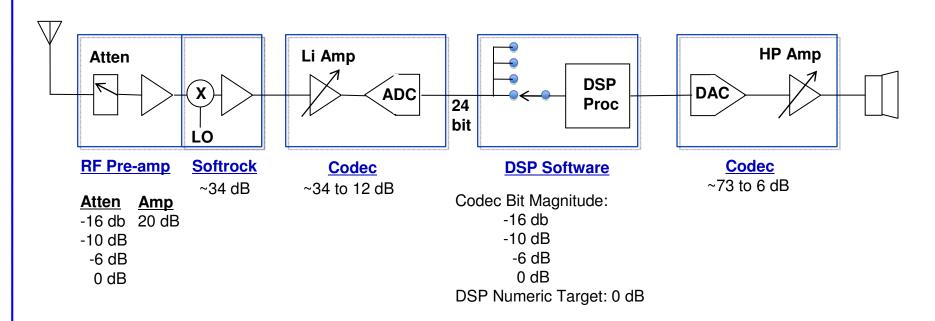


The TLV320AIC23B is a relatively old codec. The main reason we used this particular part was the accumulated SW experiences how to control it. The TLV320AIC23B also provides handy adjustable input amplifiers, mic amplifier and head phone amplifiers. Most part of the audio signal routing is done with the codec mux block and codec driver sw. Only the PSK modem audio needed external switching.

The TLV320AIC23B provides 24-bit ADC output. Our processing happens in 16-bit resolution but we can benefit from the 24 bits by selecting the magnitude that we use from it. If we start from bit 22 instead of bit 23 we get 6dB gain and so on. S/N ratio gets worse on every bit reduction. Practical numbers are 1 to 3 bits. More can be used, yielding 6 to 18dB of "free gain".



# **System Gain Distribution**



Max system gain is 72 to 90 dB depending codec output magnitude select



## **Basic System Timing**

- No operating system ... simple master loop performing computations and I/O data movement as needed.
- DSP calculations are done on every sample 1/8000Hz = 125 us sample rate
- Filter calculations for each sample takes about 16-to-20us
- Bandscope FFT is calculated in main loop when the FFT data buffer is full (16 ms interval)
- Keyer and other time critical tasks operate with 1 ms interrupt
- Processor PWM generator hardware is used for clean UI tone generation



## **Real-Time Bandscope**

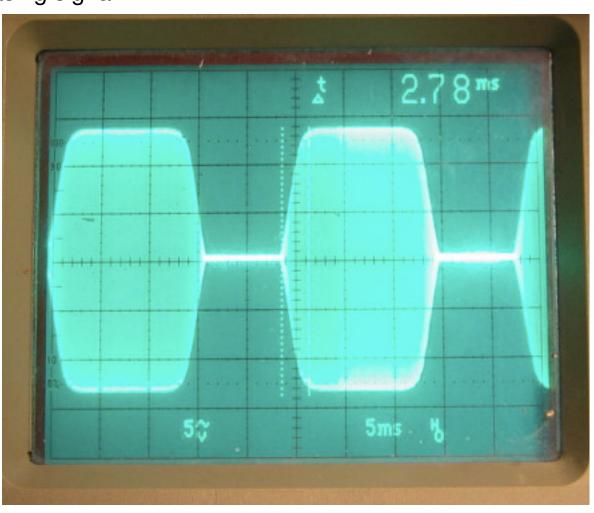


- Menu-selected FFT algorithm: Hanning, Blackman or Rectangular
- FFT is 128-sample complex FFT
- FFT is calculated with the (corrected) Microchip DSP library functions
- We need the log() function for the display bandscope. This is our only execution-optimized function. Whole process takes only 650 us.
- Calculating FFT from the complex data gives us positive and negative frequencies.
- FFT results available for LO + 4000 Hz and also LO 4000 Hz. Translates to real world application of having +/- 4000 Hz band scope.
- 4000 Hz comes from sampling frequency 8000/2
- LO frequency (display frequency) is at the center, translating to 0 Hz audio.
- USB is to the right side of the center point. LO + 4000Hz at the right corner.
- LSB is at the left side of the center point. LO -4000Hz at the left corner.
- Display marker dots shown at 1 kHz points.
- Fast attack and slow release of the display filter makes it easy to visually follow signals.



# State of the Art CW Waveform Shaping

Sinc function applied to keying envelope on leading and trailing edges reduce energy at the transitions (clicks) and provide for a pleasing signal.





## **IMPORTANCE OF I & Q GAIN MATCHING**

Multipliers are used for I and Q signals. Before correction, opposite sideband rejection was around 21 dB. After finding the optimal gain corrections, opposite sideband rejection improved to 25 dB. Gain factors for this "sweet spot" were: Left = 32768, Right = 28500. Ratio between the multipliers is 1.14975.

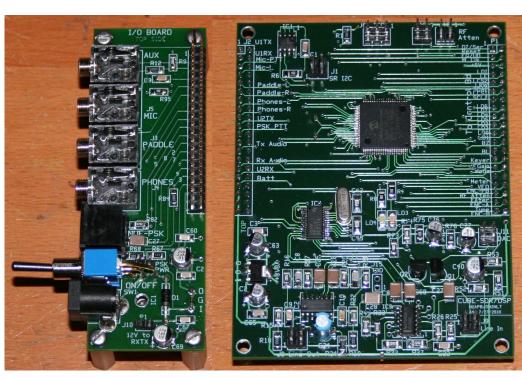
SoftRock IQ amplitude balance errors were 0.9607 V and 1.0688 V (ratio: 1.1125), so we believe the ratio magnitudes correlate quite well.

Why isn't the opposite sideband rejection better? The shortfall is phase error and general SoftRock "features". I/Q phase balance would be easy correct if a DDS is used the LO but with Si570 implementation it is not possible. (We have an option to use either the onboard/offboard Si570, or an onboard DDS.)

Small phase error correction between I and Q can be done by adding small amount I signal to Q signal or vice versa. This slight "cross mixing" of I and Q yielded good results, and rejection was improved better than 50 dB.



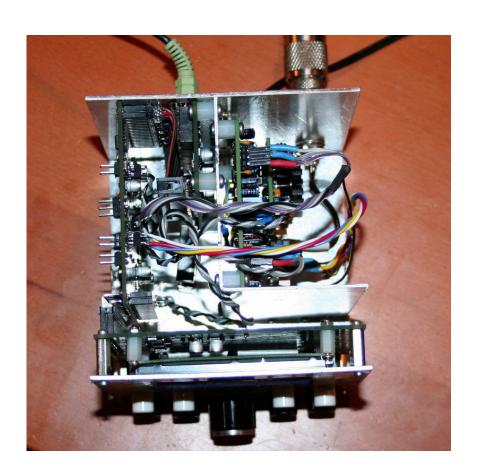
# **3-PCB BOARD SET**

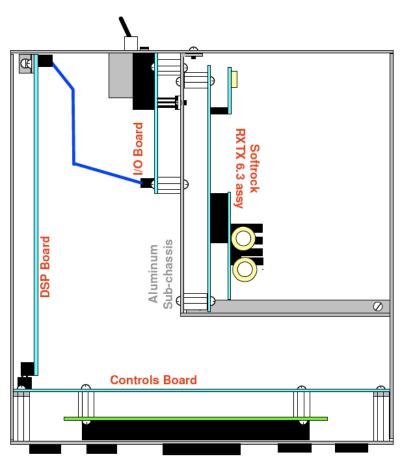






# **ENCLOSURE (TOP VIEW)**





Cube boards isolated from Softrock Transceiver by aluminum sub-chassis.

Audio & control cables between DSP and Softrock boards go through holes in sub-chassis.



# **DSP BOARD**

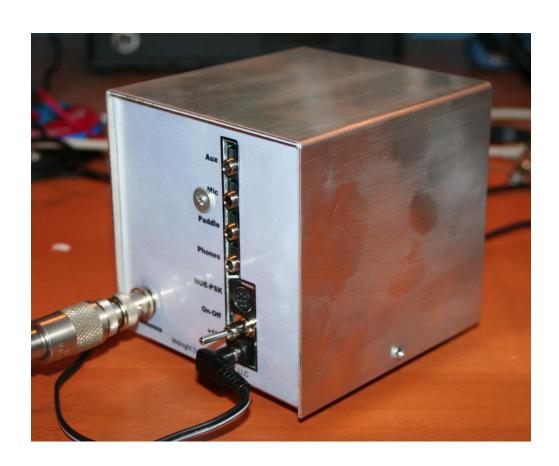
Connects I/O Board (left/behind) with Controls Board (right)





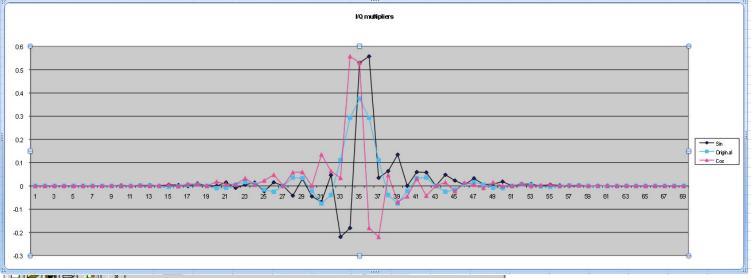
# **ENCLOSURE - REAR**

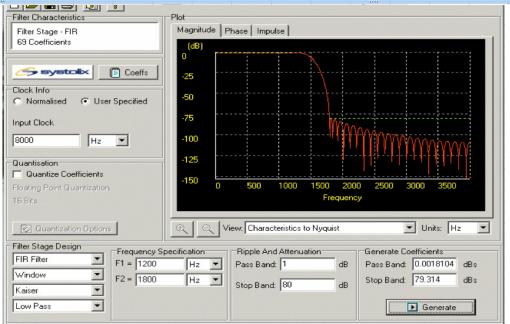
Cabling connectors & Antenna



SDR Cube

# Hilbert Filter SIN/COS pair calculator Base filter 69-tap LPF generated



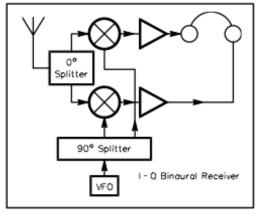


In the SDR Cube, the required 90 degree base band signal phase shifting is accomplished with a sin/cos filter pair. The required base coefficients can be calculated with any filter generator program. Then the +/- 45 deg phase shifting filters are generated using the base filter coefficients.

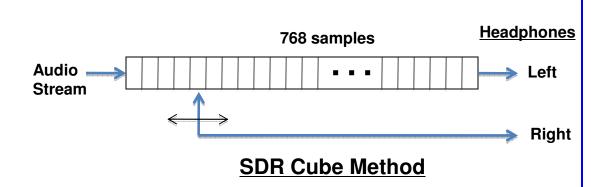


#### **BINAURAL AUDIO**

"Quasi-Binaural audio", using a delay line. In the SDR Cube we have a 768-word delay line. Samples are shifted in and out 125 us rate (8 kHz). Maximum binaural delay is 768 \* 125us = 96ms. SDR Cube binaural delay is adjustable from the menu in 1 ms steps.



"Conventional" Method



"Once my ears got used to the effect, they had to drag me away from this radio. This is one I gotta have!"—Ed Hare, W1RFI, ARRL Lab Supervisor .... QST March 1999, Rick Campbell KK7B artcle: "A Binaural I-Q Receiver"

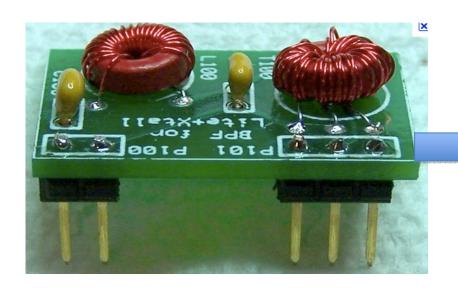


# First Cube-PSK QSO: August 26

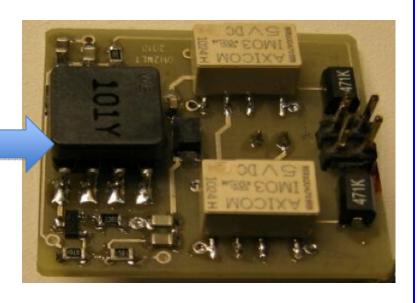




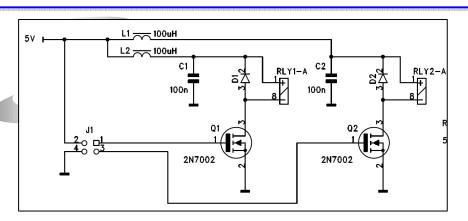
# Redesigned BPF (& RF Amp) for Softrock



Original Softrock BFP plug-in module consists of simple tuned L/C filter.

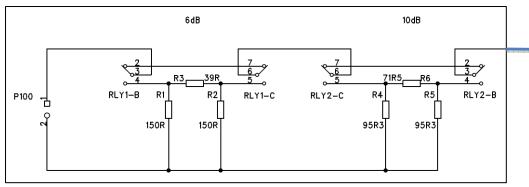


For the SDR Cube, we designed a pincompatible module that improved on the filter characteristics, as well as adding two controllable attenuator pads and an RF amp. Using this small board provides for better isolation of the antenna and more controllable gain at this important stage in the receiver.



# RF Amp & BFP

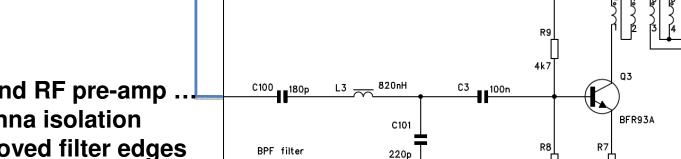
## **Relay control for RF Attenuation**



**Attenuation Options ...** 0, 6, 10, 16 dB

33R

P101



5V 39R R10

Component values for

14MHz filter

BPF and RF pre-amp ...

- antenna isolation
- improved filter edges
- 20 dB gain



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